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Non-Binary LDPC Erasure Codes with Separated Low-Degree Variable Nodes

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Knowledge for Tomorrow

Motivation (I)

- Error correcting codes are nowadays a fundamental component of modern communication networks.
- Coding at the upper-layers of the communication protocol is a simple technique to cope with packet losses.
- Applications of packet-level coding (erasure coding) within SATCOM:
 - Multicasting/broadcasting in land mobile satellite services: cope with long fading events (DVB-SH, DVB-RCS2).
 - Telemetry services in deep-space communication: reduce the average delay.
 - ► Free-space optical communication: compensate turbulence.



Principle of Packet-Level Coding

- k source packets (L bits), n encoded packets (L bits).
- (n, k) code on \mathbb{F}_q .
- CRC and error correcting code at physical layer.
- Erasures: packets whose CRC has failed after physical layer decoding.
- PEC: a packet is either correctly received or lost (erased).





Motivation (II)

- Typical erasure codes: binary low-density parity-check (LDPC) codes, Reed-Solomon codes, fountain codes (rate-less).
- Binary LDPC codes:
 - Poor performances for short codeword lengths.
 - Low decoding complexity $\mathcal{O}(n)$.
- RS codes:
 - Good performances for short codeword lengths.
 - Decoding complexity higher than $\mathcal{O}(n)$.
- Non-binary LDPC codes: good performances for short codeword lengths (AWGN).
- Non-binary LDPC erasure codes can be a flexible solution to bridge:
 - Good performances for short codeword lengths.
 - Low decoding complexity.



Outline



2 Ensemble with Separated Variable Nodes

3 Weight Distribution and Its Growth Rate

4 Code Design for the *q*-ary Erasure Channel

5 Conclusion



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Non-Binary Low-Density Parity-Check Codes

• Parity-check matrix:

$$\mathbf{H} = \begin{bmatrix} 1 & \alpha & 0 & \alpha^2 & \alpha & 0 & 0 \\ \alpha^2 & 0 & \alpha^2 & \alpha & 0 & 1 & 0 \\ 0 & 1 & 1 & \alpha & 0 & 0 & 1 \end{bmatrix}$$

• Tanner graph:



• Degree distribution pair: $\lambda(x) = \sum_{i=1}^{d_{v,\max}} \lambda_i x^{i-1}, \, \rho(x) = \sum_{i=1}^{d_c} \rho_i x^{i-1}$ $\lambda_i, \, \rho_i: \text{ fractions of edges connected to degree-} i \text{ VNs, CNs.}$

Non-Binary Unstructured LDPC Code Ensembles



- Usually, we consider sets, or *ensembles*, of LDPC codes, fulfilling $(\lambda(x), \rho(x))$.
- The design rate of the ensemble is $R = 1 \frac{\int_0^1 \rho(x) dx}{\int_0^1 \lambda(x) dx}$.
- Non-binary *unstructured* ensemble: all possible edge labelings from $\mathbb{F}_q \setminus \{0\}$ (uniform probability) and all possible edge permutations Π .



Structured LDPC Code Ensembles

- If not all edge permutations are allowed: structured ensemble.
- We focus on a structured LDPC code ensemble.
- A similar ensemble was heuristically introduced by MacKay (binary) [1].
- An ensemble similar to the one of MacKay was analyzed by C. Di (binary) [2].
- We extend the ensemble of MacKay, we provide an analytical analysis of the extended ensemble on non-binary Galois fields.
- This is the ensemble from which the progressive edge-growth (PEG) algorithm picks the codes.



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• Some of the degree-3 VNs are separated (type-e₃ edges).



• Constant CN degree d_c.





 All possible type-e₁ (brown) edge permutations Π and all possible edge labelings from F_q \ {0} (uniform probability).



Ensemble with Separated Variable Nodes: Notation



- *n*: number of VNs (codeword length in symbols from \mathbb{F}_q).
- m: number of CNs.
- V_2 : number of degree-2 VNs (type γ_2).
- V_3^S : number of separated degree-3 VNs (type γ_3).
- \tilde{V}_j : number of degree-*j* VNs of type γ_1 (brown).



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Codeword Weight Distribution

• The weight of a codeword is the number of its non-zero symbols.

Theorem 1 - $\mathbb{E}[A(\mathcal{C}, I)]$

The expected number of codewords of weight *I* for a code *C* picked from an ensemble with separated VNs (SVN ensemble) and distribution pair $(\lambda(x), \rho(x) = x^{d_c-1})$ is

$$\mathbb{E}[A(\mathcal{C}, I)] = \sum_{I: I_{\gamma_2} + I_{\gamma_3} + \sum_j \tilde{l}_j = I} {\binom{V_2}{I_{\gamma_2}}} {\binom{V_3}{I_{\gamma_3}}} \prod_j {\binom{\tilde{V}_j}{\tilde{l}_j}} \\ \times \frac{\operatorname{Coeff}\left((N^-(z))^{2I_{\gamma_2} + 3I_{\gamma_3}} (N^+(z))^{m-2I_{\gamma_2} - 3I_{\gamma_3}}, z^{\sum_j \tilde{l}_j j} \right)}{(q-1)^{-(I_{\gamma_2} + I_{\gamma_3} + \sum_j \tilde{l}_j)} {\binom{m(d_c-1)}{\sum_j \tilde{l}_j j}} (q-1)^{\sum_j \tilde{l}_j j + 2I_{\gamma_2} + 3I_{\gamma_3}}}$$

with $I = (\tilde{l}_3, \ldots, \tilde{l}_{d_{\nu,max}}, l_{\gamma_2}, l_{\gamma_3})$ and $0 \le l_{\gamma_2} \le V_2, 0 \le l_{\gamma_3} \le V_3^S, 0 \le \tilde{l}_j \le \tilde{V}_j$. Further, $N^+(z)$ and $N^-(z)$ are univariate polynomials.



Expected Block Error Probability of a q-ary LDPC Code [3]

• E.g.: (81,27) structured vs. unstructured 4-ary LDPC codes, 4-ary EC.





Growth Rate of the Weight Distribution $(n \rightarrow \infty)$

- Normalized codeword weight: $\omega = I/n$. Thus, $0 \le \omega \le 1$.
- Growth rate: $G(\omega) = \lim_{n \to \infty} \frac{1}{n} \ln \mathbb{E}[A(\mathcal{C}, \lfloor \omega n \rfloor)]$

Theorem 2 - $G(\omega)$

For an SVN ensemble with distribution pair $(\lambda(x), \rho(x) = x^{d_c-1})$ the growth rate is

$$G(\omega) = \sum_{j=3}^{d_{v,\max}} \tilde{\delta}_j \ln(B^{(j)}(x_0, y_{0,1})) + \sum_{i=2}^3 \delta_i \ln(B^{(i)}(x_0, y_{0,s})) \\ -\omega \ln(x_0) + (1-R) \ln(N^+(z_0)) + \frac{\ln(1-\beta_1 t)}{t}$$

with $\tilde{\delta}_j = \tilde{V}_j/n$, $\delta_2 = V_2/n$, $\delta_3 = V_3^S/n$, $B^{(j)}(x, y) = 1 + (q - 1)xy^j$ and $t = \frac{1}{(1-R)(d_c-1)}$. Further, $x_0, y_{0,1}, y_{0,s}, z_0, \beta_1$ are the unique solutions to a 5 × 5 system of polynomial equations.

Example of Growth Rate Curve on \mathbb{F}_4

- $\lambda(x) = \frac{1}{5}x + \frac{4}{5}x^3$, $\rho(x) = x^4$.
- Typical minimum distance:

$$\omega^* = \inf\{\omega > 0 : G(\omega) \ge 0\}.$$

- Good growth rate behaviour.
 - Large typical minimum distance.
 - Negative $G(\omega)$ for small ω .
- As $n \to \infty$,

 $\mathbb{E}(A(\mathcal{C}, \lfloor \omega n \rfloor)) \to \exp\{nG(\omega)\}.$





Growth Rate for Small (Normalized) Weight ω

Theorem 3 - $G(\omega)$ as $\omega \rightarrow 0$

The weight spectral shape of an SVN ensemble with distribution pair $(\lambda(x), \rho(x) = x^{d_c-1})$ fulfills

$$G(\omega) = -rac{3\xi_1\omega}{2} - \omega \ln\left(rac{2(1-\xi_1)(\mathrm{d}_c-2)}{
u_2(\mathrm{d}_c-1)(5\xi_1-2)}
ight) + O(\omega)$$

with $\xi_1 = \frac{2}{5} + o(1)$ and $0 < \nu_2 \le 1$.

- For small values of ω , $G(\omega)$ is always negative.
- The SVN ensemble is always characterized by a strictly positive typical minimum distance.



Typical Minimum Distances

• SVN ensemble vs. its unstructured counterpart:

$$\rho_2(x) = x^7$$

 $\lambda_2(x) = 0.1250x + 0.4951x^2 + 0.0254x^{12} + 0.2489x^{16} + 0.1056x^{17}.$

• Typical minimum distances of the two ensembles:

Ensembles	\mathbb{F}_2	\mathbb{F}_4	𝔽16	\mathbb{F}_{64}	\mathbb{F}_{128}	\mathbb{F}_{256}
SVN	0.0082	0.0178	0.0346	0.0408	0.0400	0.0373
Unstructured	0.0009	0.0017	0.0019	0.0009	0.0005	0.0003

• The SVN ensemble has much higher typical minimum distances.



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Maximum A-Posteriori Decoding

- Codeword v is transmitted, e erasures are introduced by the q-EC.
- MAP decoding: solve a linear system of *m* equations in *e* unknowns

$$\boldsymbol{\mathsf{H}}_{\overline{\mathcal{I}}}\,\boldsymbol{\mathsf{v}}_{\overline{\mathcal{I}}}^{\mathsf{T}} = \boldsymbol{\mathsf{H}}_{\mathcal{I}}\,\boldsymbol{\mathsf{v}}_{\mathcal{I}}^{\mathsf{T}}$$

- $\mathbf{v}_{\overline{\mathcal{I}}}$ and $\mathbf{v}_{\mathcal{I}}$: vector of *e* erased and (n e) received codeword symbols.
- $H_{\overline{\mathcal{I}}}$ and $H_{\mathcal{I}}$: sub-matrix composed of the corresponding columns of H.
- The system can be solved with Gaussian elimination, complexity $O(n^3)$.
- The sparseness of the parity-check matrix can be exploited in order to solve the system with reduced complexity [4].



Efficient MAP Decoding for LDPC Codes (I)

- The matrix $H_{\overline{\tau}}$ is re-organized in an approximate lower triangular form.
- The codeword symbols associated with the right-most p columns: pivots.





Efficient MAP Decoding for LDPC Codes (II)

• Zeroing-out algorithm is applied (complexity $\mathcal{O}(n^2)$):



- Gaussian elimination to recover the p pivots (complexity O(p³)).
- BP decoding to recover the remaining unknowns (complexity O(n)).
- The number of pivots can be controlled with a careful code design.



BP and MAP Decoding Thresholds [5] ϵ^* $(n \to \infty)$

- $p_{\mathsf{E}}(\epsilon) \to 0, \forall \epsilon \leq \epsilon^*.$
- *p*_E(*ϵ*): average extrinsic symbol erasure probability at the output of a decoder.





Design Guidelines, under MAP Decoding

- Design a code from an ensemble with separated variable nodes (SVNs).
- The code design in two phases (asymptotic and finite-length):
 - 1 Ensemble search with asymptotic tools.
 - 2 Construct finite-length parity-check matrix with girth optimization techniques.
- In practice:
 - 1 Search for SVN ensembles (degree distribution) with:
 - * MAP thresholds approaching the Shannon limit.
 - * BP thresholds close to the MAP threshold [6].
 - 2 Construct the finite-length parity-check matrix with PEG algorithm.



Code Performance, 4-EC (MAP decoding)

- Rate-1/2 SVN ensemble with $\overline{\epsilon}^*_{MAP} = 0.4971$, $\epsilon^*_{BP} = 0.4708$.
- Short 4-ary (256, 128) LDPC code. *n* = 256 symbols of 𝔽₄.





Decoding Complexity (MAP decoding)

• Average number of pivots, (256, 128) code, n = 256:

Ensembles	$\epsilon = 0.48$	$\epsilon = 0.46$	$\epsilon = 0.44$	$\epsilon = 0.42$	$\epsilon = 0.40$
SVN	6.96	5.01	3.35	1.74	0.75
Regular	22.95	18.99	15.59	11.62	7.62

- A (256, 128) regular ($d_v = 4$, $d_c = 8$) code has been designed.
- The asymptotic thresholds of the regular ensemble are:
 - ē^{*}_{MAP} = 0.4977.
 e^{*}_{BP} = 0.3834.
- The code from (irregular) SVN ensemble has much less pivots than the one from regular ensemble: less complexity.



Decoding Speed on the Packet Erasure Channel

• (256, 128) code on \mathbb{F}_4 over the PEC. n = 256 packets of 1024 bytes.





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Conclusion

- The design of non-binary LDPC erasure codes has been investigated.
- A promising ensemble of LDPC codes has been identified and analyzed in terms of:
 - Asymptotic thresholds.
 - Weight distribution.
 - Growth rate of the weight distribution.
- Codes from the ensemble designed and analyzed in terms of:
 - Performance (codeword error rate).
 - Decoding complexity.
- Codes from the ensemble provide excellent trade-offs between:
 - ► Waterfall performance, error-floor and decoding complexity.
- Thanks to their flexibility they can be used in many practical applications.



G. Garrammone, E. Paolini, B. Matuz, G. Liva, "*Non-Binary LDPC Erasure Codes with Separated Low-Degree Variable Nodes*", IEEE Transactions on Communications, submitted.

Thank you for your attention! Questions?



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References

MacKay et al., Near Shannon limit performance of low density parity check codes, Electronics Letters, vol. 32, no. 18, pp. 1645-1646, 1996.

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- ② C. Di, Asymptotic and Finite-Length Analysis of Low-Density Parity-Check Codes. PhD Thesis, E.P.F.L. Press, 2004.
- O Liva et al., Bounds on the error probability of block codes over the q-ary erasure channel, IEEE Trans. Commun., vol. 61, no. 6, pp. 2156-2165, Jun. 2013.
- Observe and the server of t
- S Ashikhmin et al., Extrinsic information transfer functions: Model and erasure channel properties, IEEE Trans. Inform. Theory, vol. 50, no. 11, pp. 2657-2673, Nov. 2004.
- Measson et al., Maxwell construction: The hidden bridge between iterative and maximum a posteriori decoding, IEEE Trans. Inform. Theory, vol. 54, no. 12, pp. 5277-5307, Dec. 2008.

